

METHOD AND APPARATUS FOR ANALYZING NOISE IN A DIGITAL  
CIRCUIT, AND STORAGE MEDIUM STORING AN ANALYSIS PROGRAM

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BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates to a method and an  
apparatus for analyzing noise in a digital circuit. In  
particular, the present invention is concerned with a method  
and apparatus for analyzing noise generated in a digital  
circuit incorporating digital ICs, and also to a storage  
medium storing an analysis program.

2. Description of the Related Art

In digital circuits incorporating digital ICs, filters  
are commonly used to exclude noise generated within the  
digital circuits, which may affect other electronic devices.  
It has not been easy, however, to select a filter, since the  
filter must match both the particular digital ICs, and the  
circuit boards on which the digital ICs are mounted, having  
different characteristics from one another. Accordingly, in  
order to analyze noise in a digital circuit and to thereby  
facilitate selection of an appropriate noise filter, a  
method has been proposed in which an input signal of a  
digital circuit is divided in the time domain and a

corresponding output signal is computed for each of the divided time steps.

Although the conventional method allows precise analysis, the method requires computations for a lot of time steps before reaching a steady state, and also requires complex and sophisticated computations to obtain the relevant characteristics of the digital ICs. Therefore, the method requires the use of a high-end computer and expensive software, raising the relevant cost. On the other hand, if an ordinary personal computer were used, the analysis would take too long.

#### SUMMARY OF THE INVENTION

Accordingly, the present invention provides a method of analyzing noise in a digital circuit by performing a simulation, which enables quick analysis using an ordinary personal computer running reasonably priced software.

The present invention also provides an apparatus for analyzing noise in a digital circuit, which employs software to perform noise simulations with ease.

The present invention further provides a storage medium storing such software.

To these ends, the present invention in one aspect provides a method for analyzing noise in a digital circuit of the type that has a digital IC, a passive circuit, and a

transmission line. The method includes the steps of  
obtaining the transfer function of the digital circuit from  
the circuit constants of an equivalent circuit of the  
digital IC constituted of passive components, the circuit  
5 constants of the passive circuit, and the circuit constants  
of the transmission line; expanding an input signal to the  
digital circuit into a sine-wave series by Fourier  
expansion; obtaining a frequency-domain output spectrum of  
the digital circuit from the transfer function of the  
10 digital circuit and the sine-wave series; and transforming  
the frequency-domain output spectrum into a time-domain  
output waveform by inverse Fourier transformation.

The present invention in another aspect provides an  
apparatus for analyzing noise in a digital circuit of the  
type that incorporates a digital IC, a passive circuit, and  
5 a transmission line. The apparatus includes a first input  
device for inputting the circuit constants of an equivalent  
circuit of the digital IC constituted of passive components;  
a second input device for inputting the circuit constants of  
the passive circuit; a third input device for inputting the  
20 circuit constants of the transmission line; a first  
operation unit for obtaining the transfer function of the  
digital circuit from the circuit constants of the equivalent  
circuit of the digital IC constituted of passive components,  
25 the circuit constants of the passive circuit, and the

circuit constants of the transmission line; a second operation unit for expanding an input signal to the digital circuit into a series of sine waves by Fourier expansion; a third operation unit for obtaining a frequency-domain output spectrum of the digital circuit from the transfer function of the digital circuit and the series of sine waves; and a fourth operation unit for transforming the frequency-domain output spectrum into a time-domain output waveform by inverse Fourier transformation.

In accordance with the present invention, the Fourier expansion of an input signal permits the latter to be expressed in terms of a sum of a series of sine waves. Transformation or reduction of a digital IC, which is a non-linear component, into an equivalent circuit composed of passive components permits a circuit analysis by a simple linear computation, without requiring complicated non-linear computations to determine the characteristics of an IC. More specifically, the transfer function of a digital circuit can be determined based on the circuit constants of the digital IC equivalent circuit and on the circuit constants of the passive components and the transmission line. It is possible to obtain a frequency-domain output spectrum, based on the transfer function and the Fourier-expanded input signal. It is therefore possible to obtain a

time-domain output waveform, through an inverse Fourier transformation of the frequency-domain output spectrum.

The noise analysis method of the present invention may be implemented by a program which performs noise analysis in accordance with the described procedure. The transfer function of the digital circuit can be determined simply by inputting the circuit constants. This enables computation of output signals for a variety of input signals and, hence, for a variety of classes of components. This makes it possible to select an optimum filter, based on an output waveform obtained when such a filter is incorporated in the digital circuit.

The apparatus may include a first storing device for storing the circuit constants for respective equivalent circuits of a plurality of digital ICs, whereby the circuit constants of a digital IC selected from the plurality of digital ICs may be read out from the first storing device and input to the first operation unit. This eliminates the need to enter the circuit constants individually.

The apparatus may also include a calculation device for calculating the circuit constants of the transmission line based on factors including the line width and line length of a pattern formed on a wiring board, circuit board thickness, and substrate material, whereby the calculated circuit constants may be input to the first operation unit. This

allows the user to enter the characteristic values instead of calculating the circuit constants manually.

The apparatus may also include a second storing device for storing the circuit constants of a plurality of passive circuits, whereby the circuit constants of a passive circuit selected from the plurality of passive circuits may be read out from the second storing device to be input to the first operation unit. By storing the circuit constants of, for example, noise filters, the effects of the filters can be simulated without entering the circuit constants of the filters individually.

Preferably, the apparatus includes a display for displaying passive circuit characteristics when selecting one of the plurality of passive circuits, so that the passive circuit characteristics information is available when selecting a desired passive circuit.

It is preferable that the frequency-domain output spectrum obtained by the third operation unit and the time-domain output waveform obtained by the fourth operation unit are displayed on the display, thus enabling confirmation of the results of the analysis.

It is also preferable that the impedance-frequency characteristics of the digital circuit are displayed on the display.

It is also preferable that the results of operations performed with different transfer functions are simultaneously displayed on the display, so that the analysis results can readily be compared.

5       The device may also include a printing device for simultaneously printing input information regarding the transfer function of the digital circuit, and the results of simulation, so that the analysis results can be reviewed in association with the conditions of the analysis.

10       The present invention, in still another aspect thereof, provides a storage medium storing a program. The program includes the steps of obtaining the transfer function of the digital circuit from the circuit constants of an equivalent circuit of the digital IC constituted of passive components, the circuit constants of the passive circuit, and the  
15       circuit constants of the transmission line; expanding an input signal to the digital circuit into a series of sine waves by Fourier expansion; obtaining a frequency-domain output spectrum of the digital circuit from the transfer  
20       function of the digital circuit and the series of sine waves; and transforming the frequency-domain output spectrum into a time-domain output waveform by inverse Fourier transform. By installing the program on a computer from the storage medium, the computer can be used as a noise analyzer  
25       for digital circuits.

The invention further relates to a method of enabling a user to select a passive circuit, the passive circuit being included in a digital circuit along with a digital IC and a transmission line, the method comprising the steps of:

5 supplying the user with a program for analyzing noise in the digital circuit, on the basis of circuit constants of the digital IC, the transmission line, and the passive circuit; and supplying the user with circuit constants of a plurality of passive circuits, thereby enabling the user to select a  
10 passive circuit from among the plurality of passive circuits on the basis of an analysis result of said program.

The above and other features and advantages of the present invention will become apparent from the following description of embodiments of the invention taken in  
15 conjunction with the accompanying drawings.

#### BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 shows an initial screen which appears upon activation of an apparatus for analyzing noise in a digital  
20 circuit according to an embodiment of the present invention;

Fig. 2 shows a screen prompting a user to enter circuit constants of a transmitter IC and a receiver IC;

Fig. 3 shows a window for enabling the user to select a filter;



Fig. 4 is a diagram showing a relationship between a wiring board and a transmission line fabricated thereon;

Fig. 5 shows a screen which appears when a plurality of input settings have been entered;

5 Fig. 6 shows a screen displaying analysis results; and

Fig. 7 shows a screen displaying results of a plurality of analyses performed with different settings; and

Fig. 8 shows a printout of analysis results.

#### DESCRIPTION OF EMBODIMENTS OF THE INVENTION

Fig. 1 shows an initial screen appearing on a display connected to a noise analyzing apparatus according to the present invention. By way of example, the noise analyzing apparatus may be a personal computer installed with a program for implementing a noise analyzing method according to the present invention. When the program is activated, the initial screen as shown in Fig. 1 appears. Shown at the top of the screen is a circuit diagram of a simulation circuit 10. The noise analyzing apparatus is used for  
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analyzing noise in a digital circuit incorporating digital ICs. Digital ICs can be provided to perform a variety of circuit functions, such as transmitting, receiving, amplification and/or signal processing. The circuit diagram shows an input signal (1), a transmitter IC (2), a passive circuit (3), a transmission line (4), a receiver IC (5), and

a simulation point, which is the location in the simulated circuit where the simulated output wave form exists. In the illustrated embodiment, the passive circuit is a noise filter, and the transmission line is a wiring board on which the digital ICs are mounted.

The upper left of the screen is an input signal specifying section 12, in which the input signal is specified in terms of clock frequency, duty ratio, and rise time and fall time. These characteristic values of the input signal are entered via an input device, for example, a keyboard.

In the lower left of the screen is a transmitter IC specifying section 14. The transmitter IC is either selected from a list of ICs via an input device, for example, a mouse, or specified by entering IC values of inductance, capacitance, and resistance via an input device, for example, a keyboard.

For the settings in the transmitter IC specifying section 14, a variety of digital ICs have been transformed into the form of equivalent LCR circuits composed of passive elements. That is, the values of inductance, capacitance, and resistance for each of a plurality of LCR circuits transformed from a variety of digital ICs are stored in a storage medium such as a memory or a hard disk. Thus, once a digital IC to be used is selected from the list, the

values of inductance, capacitance, and resistance for an equivalent LCR circuit of the digital IC are automatically read from the storage medium to be entered in the corresponding boxes.

5           In selecting a digital IC, initially a choice is made between CMOS IC and TTL IC in the IC type box. When CMOS IC is chosen, product codes of CMOS ICs are displayed in the product code box, so that the user can select one of the CMOS ICs. When TTL IC is chosen, one of the TTL ICs is selected in a similar manner.

10           When the user has chosen on the transmitter IC setting section 14 to enter the output characteristics of an IC in terms of LCR, an LCR circuit transformed from the digital IC is displayed as shown in Fig. 2, thus permitting values of the inductance, capacitance and resistance to be entered individually.

15           At the center of the screen is a filter specifying section 16 for specifying a filter, which is a passive circuit, in which a filter is selected from a list of filters stored in a storage medium. As shown in Fig. 3, a window 18 for enabling selection of a filter is activated. In the window 18, the user selects the filter type such as the chip ferrite bead type, the filter size, and the filter impedance, in that order. The window 18 includes boxes for  
20           macro-, medium-, and micro-classifications, as well as a  
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product code box. The macro-classification box is used to enable the user to select the type of the desired filter, e.g., a chip-ferrite bead filter. The medium-classification box is used for selection of the size of the filter of the type specified in the macro-classification box, while the micro-classification box permits selection of the impedance of the selected filter. When a plurality of filters match the specified characteristics, the impedance-frequency characteristics for each of the plurality of filters are displayed, together with the product code of the filter. Then, a desired filter is selected by examining the characteristics of the plurality of filters.

In the lower center of the screen is a transmission line specifying section 20, which permits settings of the characteristic impedance of a wiring board on which the digital ICs are to be mounted. The characteristic impedance is calculated from the line length and line width of a pattern fabricated on a substrate, and the thickness and material of the substrate. More specifically, the characteristic impedance  $Z$  and the propagation constant  $\beta$  of the transmission line are expressed, respectively, by the following equations 1 and 2, where  $l$  represents the line length.

$$Z = \frac{60}{\sqrt{\epsilon_{ef}}} \ln \left[ f \frac{h}{W_0} + \sqrt{1 + (2h / W_0)^2} \right] \quad (1)$$

$$\beta = \omega \sqrt{\epsilon_{ef}} / c_0 \quad (2)$$

The constants  $\epsilon_{ef}$ ,  $f$ , and  $W_0$  are expressed, respectively, by the following equations 3, 4, and 5. As shown in Fig. 4,  $h$  is the thickness of the substrate,  $W$  is the electrode width,  $t$  is the electrode thickness,  $\epsilon_r$  is the relative dielectric constant of the substrate, and  $c_0$  is the velocity of light in vacuum.

$$\epsilon_{ef} = \frac{\epsilon_r + 1}{2} + \frac{\epsilon_r - 1}{2} (1 + 10h / W)^{-1/2} \quad (3)$$

$$f = 6 + 0.283 \exp [-(30.7h/W_0)^{0.753}] \quad (4)$$

$$W_0 = W + \frac{t}{\pi} \ln \frac{4e}{\sqrt{(t / h)^2 + \frac{1}{\pi^2(W / t + 1.1)^2}}} \quad (5)$$

As described, by incorporating a program for calculating the characteristic impedance of the transmission line from the line length, line width, substrate thickness, and substrate material, the characteristic impedance can readily be specified and entered.

In addition, there may be a user-selectable mode for enabling manual entry of a characteristic impedance which for example may be calculated by the user.

Further, it is also possible to arrange for the transmission line to be disregarded. In such a case, the noise analysis is performed on the assumption that there is no impedance component attributable to the transmission line.

On the right side of the screen is a receiver IC specifying section 22, in which IC type and IC product code can be selected in a similar manner as in the transmitter IC specifying section 14. When the user chooses to specify input characteristics by entering the values of inductance, capacitance, and resistance, an LCR circuit is displayed with boxes to enter those values individually, as shown in Fig. 2.

When the settings are complete as shown in Fig. 5, the "Start simulation" button is clicked on to start the noise analysis. First, the transfer function of the digital circuit is calculated. A two-terminal circuit component can

be described by a 2x2 admittance matrix  $y$  as expressed in the following equation (6).

$$y = \begin{bmatrix} Y_{11} & Y_{12} \\ Y_{21} & Y_{22} \end{bmatrix}$$

5 (6)

The admittance matrices for all the circuit components are summed. The following equation (7) expresses an example where a component A is connected between node 2 and node 3, and a component B is connected between node 3 and node 4 (junction points between the above-described circuit components are shown in Fig. 1 as nodes 1, 2, 3, 4, 5 and 6).

$$\begin{bmatrix} 0 & 0 & 0 & 0 & \cdots \\ 0 & Y_{11}^A & Y_{12}^A & 0 & \cdots \\ 0 & Y_{21}^A & Y_{22}^A + Y_{11}^B & Y_{12}^B & \cdots \\ 0 & 0 & Y_{21}^B & Y_{22}^B & \cdots \\ \vdots & \vdots & \vdots & \vdots & \ddots \end{bmatrix} \begin{bmatrix} V_1 \\ V_2 \\ V_3 \\ V_4 \\ \vdots \end{bmatrix} = \begin{bmatrix} I_1 \\ I_2 \\ I_3 \\ I_4 \\ \vdots \end{bmatrix}$$

(7)

The transfer function of the simulation circuit 10 is obtained by summing the admittance matrices as above based on the settings of the components for all the components included in the simulation circuit 10. The admittance matrices of a resistor, a capacitor, an inductor, a

transmission line, and a custom component, are expressed by,  
respectively, equations (8), (9), (10), (11), and (12). In  
equation (11), the characteristic impedance  $Z$  is obtained by  
equation (1), and the propagation constant  $\beta$  is obtained by  
equation (2). The custom component expressed by equation  
(12) is a digital IC herein.

$$Y = \begin{bmatrix} 1/R & -1/R \\ -1/R & 1/R \end{bmatrix} \quad (8)$$

$$Y = \begin{bmatrix} j\omega C & -j\omega C \\ -j\omega C & j\omega C \end{bmatrix} \quad (9)$$

$$Y = \begin{bmatrix} 1/j\omega L & -1/j\omega L \\ -1/j\omega L & 1/j\omega L \end{bmatrix} \quad (10)$$

$$Y = \frac{1}{jZ \sin \beta l} \begin{bmatrix} \cos \beta l & -1 \\ -1 & \cos \beta l \end{bmatrix} \quad (11)$$

$$Y = \begin{bmatrix} 1/Z(\omega) & -1/Z(\omega) \\ -1/Z(\omega) & 1/Z(\omega) \end{bmatrix} \quad (12)$$



When the transfer function of the simulation circuit 10  
is obtained, then the input signal is Fourier-expanded.  
That is, the time-domain input signal is reduced to a  
frequency-domain sine wave series, thereby enabling  
5 frequency response analysis. For instance, if  $V_{in}(t)$ , an  
input voltage at time  $t$ , cyclically varies with a period  $T$ ,  
 $V_{in}(t)$  can be approximated by a series of  $N$  sine waves as  
expressed in the following equations (13).

$$V_{in}(t) \cong \sum_{n=0}^N c_n \exp(j\omega_n t), \quad \omega_n \equiv 2\pi n / T$$

(13)

In the above equation (13),  $C_n$  is the magnitude of the  
angular frequency  $\omega_n$  component of the input signal. The  
15 output voltage  $F[c_n \exp(j\omega_n t)]$  corresponding to the component  
can readily be obtained by an ordinary circuit calculation  
using the previously obtained transfer function of the  
digital circuit. The amplitude  $F_n$  is expressed by the  
following equation (14).

$$F[c_n \exp(j\omega_n t)] = F_n \exp(j\omega_n t)$$

(14)

From the above equation (14), a voltage spectrum represented in the frequency domain is obtained. Then, the frequency-domain voltage spectrum is transformed into a time-domain output voltage  $V_{out}(t)$  waveform by inverse Fourier transform, as expressed by the following equation (15).

$$V_{out}(t) = \sum_{n=0}^N F_n \exp(j\omega_n t) \quad (15)$$

The output voltage thus obtained as the analysis result is expressed on the screen by the frequency-domain voltage spectrum and the time-domain voltage waveform, as shown in Fig. 6. Furthermore, the output voltage waveform is simulated and displayed both for the case where the filter is used and the case where the filter is not used, so that the effect of the filter is well understood. Also shown on the screen are the impedance-frequency characteristics of the digital circuit. Based on the simulation results, a filter which provides desired output characteristics can be selected.

It is preferable for the results of a plurality of simulations obtained with a variety of set values to be simultaneously displayed on one screen, as shown in Fig. 7,

so that output waveforms provided by different filters can readily be compared. It is also preferable for the results to be printed out via a printer as shown in Fig. 8 so that the results can be used for later examination. Furthermore, the printout may include the simulation settings as well as the simulation results, so that the settings and the results can be correlated.

The noise analysis method as described above allows batch processing and therefore quick calculation, as opposed to the conventional method in which the input signal is divided into a large number of time steps and the calculation is repeated for the successive time steps. In addition, the described noise analysis method can be implemented by a simpler program than the conventional method, and thus can be used on an ordinary personal computer.

Such a program for implementing a noise analysis method according to the present invention may be stored on a storage medium, for example, a CD-ROM, for being installed on a personal computer. Alternatively, the program may be downloaded from a website. In either case, the personal computer on which the program is installed works as a noise analyzing apparatus according to the present invention.

Sales of filter products can be promoted by distributing the program to prospective filter customers, by

allowing downloading from a web site or by means of CD-ROMs,  
for example. To this end, the filter setting section 16 is  
arranged for selection of the filter products of the program  
distributor, such as the filter manufacturer, to enable  
5 potential purchasers to use the noise analyzing method when  
considering whether to purchase those filter products.  
Thus, the use of the analysis program on the stored data  
concerning the manufacturer's filters provides customer with  
information concerning the output waveforms obtainable with  
10 and without each such filter, which naturally drives the  
customers to purchase those filters. A purchaser also can  
enjoy the advantage of having a chance to confirm the output  
waveform before placing a purchase order.

Although the invention has been described in connection  
with embodiments thereof, it is to be understood that the  
described embodiments are only illustrative and various  
15 changes and modifications may be imparted thereto without  
departing from the scope of the present invention.